Quantum Fourier Transforms

Special Topics in Computer Science: Quantum Computing CSC591/ECE592 – Fall 2019

Outline

- Math review complex roots of unity
- Introduction concept of Fourier transforms for continuous functions
- Applying Fourier transforms to digital computers
- Discrete Fourier Transform
- Mapping Discrete Fourier Transform to Quantum Fourier Transform
- Derivation of formula for generalized Quantum Fourier Transform
- Worked example Quantum Fourier Transform for 3 qubits
- Mapping 3 qubit Quantum Fourier Transform to quantum computing gates

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Math Review - Roots of Unity

• Let
$$\omega_N \equiv e^{2\pi i/N}$$

• For a given N generate all possible values

$$\{1, \omega^1, \omega^2, \omega^3, \dots, \omega^{N-1}\}$$

{1,
$$\omega^{e^{2\pi i/N}}, \omega^{e^{4\pi i/N}}, \omega^{e^{6\pi i/N}}, \dots, \omega^{(N-1)e^{2\pi i/N}}$$
}

 Replace the continuous version of x, e^{isx} parameterized by s) into N vectors

$$\mathbf{v}_{j} = \begin{pmatrix} v_{j0} \\ v_{j1} \\ \vdots \\ v_{jk} \end{pmatrix} = \begin{pmatrix} \omega^{-j0} \\ \omega^{-j1} \\ \vdots \\ \omega^{-jk} \end{pmatrix}$$

where k is the coordinate index and j is the parameter that labels each vector

Nth Roots of Unity

• Begin with the Taylor series expansion of the exponential function

$$\omega_N = \exp\left(\frac{2\pi i}{N}\right) = \cos\left(\frac{2\pi}{N}\right) + i\sin\left(\frac{2\pi}{N}\right)$$

• Compute the unit modulus

$$|\omega_N|^2 = \omega_N^* \omega_N = \exp\left(-\frac{2\pi i}{N}\right) \exp\left(\frac{2\pi i}{N}\right) \exp\left(-\frac{2\pi i}{N} + \frac{2\pi i}{n}\right)$$
$$= \exp(0) = 1$$

• Make a graphical construction of these N roots of unity

Graphical Construction of These N Roots of Unity

The sum of all sums:

$$\frac{1}{N}\sum_{k=0}^{N-1}(\omega_N)^{jk}=\delta_{j,0}$$

Construct a coordinate system in the complex plane with real and imaginary axes



Example 8th Roots of Unity Plotted on Complex Plane





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Fourier Transforms

- FT is a mapping between two domains
 - Time and frequency
 - position and momentum
- Can combine many different signals each with their own frequency, amplitude and phase



Fourier Transform for Continuous Functions

$$f(x) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} A(k) e^{ikx} dk$$
$$A(k) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(x) e^{-ikx} dx$$

Generic Expression for Fourier Transform and Inverse Fourier Transform

$$F(s) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} f(x) e^{-isx} dx .$$

$$f(x) = \frac{1}{\sqrt{2\pi}} \int_{-\infty}^{\infty} F(s) e^{isx} ds .$$

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Application of Fourier Transform Ideas to Digital Computers

- Continuous mathematical formulations are incompatible with digital or quantum computers
- Computers are discrete and finite collections of bits (or qubits)
- Need to modify the continuous Fourier Transform to a digital formulation
- This digital formulation needs to ultimately be integrated into the hardware architecture of a quantum computer
- To make this happen it means designing and building gate operations that can be understood by a quantum computer

Discrete Fourier Transform (DFT)

- Need to construct an equivalent summation expression that can reference the Discrete Fourier Transform's functionality versus the continuous Fourier Transform
- **Definition:** An nth order DFT is a function of an n-component vector $f = (f_k)$ that produces a new n-component vector $F = (F_j)$ given by a formula that describes the output vector's n-components

$$F_j \equiv \frac{1}{\sqrt{N}} \sum_{k=0}^{N-1} f_k \omega^{-jk}$$
 j=0, 1,....N-1

- {F_j} provides the "weighting factors" for the expansion of f as a weighted sum of the frequencies ω^j $f_k = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} F_j \omega^{jk}$
- Essentially the real functions that were used in the continuous Fourier transforms have become complex roots of unity in the discrete Fourier transform

Matrix Representation of a Discrete Fourier Transform

$$\begin{array}{c} NUTATION & d \equiv \omega^{1} & AMD & \omega \geq \omega_{H} \equiv gHg\left(\frac{zH^{1}}{H}\right) \\ THIS & MATRIX ENERDES THE OFT FUNCTIONALITY \\ \hline THIS & MATRIX ENERDES THE OFT FUNCTIONALITY \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & UECTO(2 & S_{H} CAN BE WRITTEN TN MATRIX FURM \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1})} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1}(SN^{-1}(SN^{-1}))} \\ \hline THE & 1 & \dots & 1_{H^{-1$$

Consequences of Transformation from Continuous Fourier Transform to Discrete Fourier transforms

- Three consequences
 - Integrals become sums from 0 to N-1. [Rather than evaluating f at real continuous x the evaluation is over a complex spectrum vector F_j that also has N components
 - 2. The factor $\frac{1}{2\pi}$ is replaced by a normalizing factor $\frac{1}{\sqrt{N}}$

NOTE: This choice is driven by the need for all vectors to live in the projective sphere in Hilbert space and therefore be normalized

3. The complex roots of unity ω^{jk} replace the general exponential e^{isx}

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SUMMARY - DET WORKS UIX> = Z IVXVIUIX UIX> = Z IVXVIUIX DIO V INTERT ELEMENT SELON 22015 = Z UIDXIIN (SINDE ZUKIDX) XX $U \sum_{x} F(x) (x) = \sum_{x} F(x) u(x)$ $= \sum_{x} F(x) \sum_{y} k(x,y) (y)$ = Z (Z K(X,S) FLX)) 197 = Z (Z K(X,S) FLX)) 197 = Z JUNGRSE FRUIRIER TRANSFORM

Transforming from Discrete Fourier Transform to Quantum Fourier Transform

• Defined the Discrete Fourier Transform that a given vector $x \in \mathbb{C}^N$ outputs another vector $y \in \mathbb{C}^N$ such that

$$y_k = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} x_j \omega^{jk}$$

 The Quantum Fourier Transform transforms a basis set {|0>, |1>, ...{N-1>} into another basis set such that

$$|j\rangle = \frac{1}{\sqrt{N}} \sum_{j=0}^{N-1} |k\rangle \omega^{jk}$$

Matrix Representation of a Quantum Fourier Transform

$$OFT_{N} = \prod_{i=1}^{l} \begin{pmatrix} 1 & 1 & 1 & \dots & 1 \\ 1 & \omega & \omega^{2} & \omega^{3} & \cdots & \omega^{(N-1)} \\ 1 & \omega^{2} & \omega^{2} & \omega^{2} & \omega^{2} & \cdots & \omega^{(N-1)} \\ 1 & \omega^{2} & \omega^{2} & \omega^{2} & \cdots & \omega^{(N-1)(N-1)} \end{pmatrix}$$

$$= ENTRIES IN THE MATRIX ARE THE NTE RUOTS OF UNITY
$$\omega = R^{2\frac{N}{N}} = \cos\left(\frac{2\pi}{N}\right) + i \sin\left(\frac{2\pi}{N}\right)$$

$$= ENTRIES IN THE PATRIX ARE THE NTE ROOTS OF UNITY
$$\omega = R^{2\frac{N}{N}} = \cos\left(\frac{2\pi}{N}\right) + i \sin\left(\frac{2\pi}{N}\right)$$

$$= Intermediate (1+i)$$

$$\omega = Intermediate (1+i)$$

$$\omega = Intermediate (1+i)$$

$$\omega = Intermediate (1+i)$$

$$\omega = N^{-1}$$$$$$

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16

Derivation of the Recursion Relation for Nth Order QFT

Notation

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Sample Derivation for N=3

$$\frac{N=3}{N} \frac{GET}{V} = \frac{1}{2} \frac{1}{2$$

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19 November 20

21 November 20

Group Even Terms and Factor Right-Most Bit

$$\frac{N \times 3}{5} \underbrace{G_{x,AMPLE}}_{Y-EVEN} \underbrace{Y-EVEN}_{GROUPING} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPING}_{Y-EVEN} \underbrace{GROUPR}_{Y-EVEN} \underbrace{Groupr}_{Y-EVE$$

20

Group Even Terms and Factor Right-Most Bit (cont'd)

$$\frac{N=3}{CONSEQUENCES} \xrightarrow{OF} FACTURING OUT CIRCUIT - EVEN GROUPING}{CONSEQUENCES OF FACTURING OUT RICHT MOST BIT}$$

$$1) \xrightarrow{T}_{yough} \xrightarrow{J}_{out} HALVES THE FUTAL SUM$$

$$2) (Y_2 y_1 y \xrightarrow{J}_{out} Hy y_0 y_0 y)$$

$$3) \xrightarrow{T}_{1} \xrightarrow{J}_{0} \xrightarrow{T}_{1} SHIFTS THE PRODUCT FACTOR$$

$$4) \xrightarrow{2}_{R} \xrightarrow{2}_{R} \xrightarrow{2}_{R} \xrightarrow{1}_{1}$$
FOLD ABOVE 4 BULLETS INTO THE Y-EVEN PORTION OF QATEGO.
FOLD ABOVE 4 BULLETS INTO THE Y-EVEN PORTION OF QATEGO.

$$= \left[\sum_{y=v}^{3} (\prod_{k=v}^{1} (w^2)^{xyk} x^{2k}) |(y_1 y_0 y)\right] |0\rangle$$

21

Generalize to "N" Even Terms



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Group Odd Terms and Factor Right-Most Bit

ODD GROUPING 1 N=3 GXAMPLE START WITH CASE 9"=1 START WITH CASE 9"=1 W *9.2° = W *(2)(1) = W * W *9.2° = W *(2)(1) = W * FACTORING OUT W TERM ALLOWS THE PRODUCT TO R=1 C START AT R=1 THE EVEN GRO # ALIGNS WITH EVEN GROUP FURMULATION y-ODD GROWP = \$\$1 10017 + \$23 0117 + \$25 11017 + \$25 11017 + \$201117 = [\$x11007 + \$x31017 + \$x51107 + \$x71117]117 = w [J (T w SR 2) 1 y 2 317] 117 2-01717

Group Odd Terms and Factor Right-Most Bit (cont'd)

QFT CIRCUIT
y ODD GRUMPING
CONSEGURENCES OF FACTORING OUT REAT MUST BIT
1.)
$$\frac{7}{2} \rightarrow \frac{3}{2}$$
 HALVES THE TOTAL SUM
2.) $1y_2y_1 \rightarrow 1y_1y_0$?
3.) $\frac{7}{1} \rightarrow \frac{7}{2}$ SNIFTS THE PRODUCT FACTOR
4.) $2^{\frac{1}{2}} \rightarrow 2^{\frac{1}{2}+1}$
FULL THE ABOVE Y BULLEFS INTO THE Y-ODD PORTION OF THE OFT GO
FULD THE ABOVE Y BULLEFS INTO THE Y-ODD PORTION OF THE OFT GO
 $y_{-0}p_{D} = w \left[\frac{3}{2} \left(\prod_{k=0}^{1} (w^2)^{xy_k 2^k} \right) 1y_1y_0 \right] 11?$

Generalize to "N" Odd Terms



Quantum Fourier Transform Recursion Relation

$$\begin{aligned} & QFT \leq \int RCUIT \\ REPEATING THIS PROLODURG (RECURSIVELY) \\ QFT^{2^{n}} | x \rangle^{m} = \prod_{k=1}^{m} \left(\frac{107 + \omega_{2^{k}} | 17}{\sqrt{2}} \right) \\ RHS OF THE ABOVE EQ. IS EXPRESSED IN TERMS OF ROOTS UP CANITY \\ & \omega_{2} \omega_{4} = \omega_{2^{m-k}} \\ & \omega_{2^{k}} = \prod_{k=1}^{m} \frac{107 + \omega_{k}^{2^{m-k}} | 17}{\sqrt{2}} \\ & QFT^{2^{n}} | x \rangle^{n} = \prod_{k=1}^{m} \frac{107 + \omega_{k}^{2^{m-k}} | 17}{\sqrt{2}} \end{aligned}$$

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Example of a 3 Qubit Quantum Fourier Transform

8th Complex Roots of Unity



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General Formula for QFT and First Factor for the 3 Qubit Example Construction

TRANSLATING THE MATH TO A OFT GUANTUM CHRUIT
GRAMPLE M=3
OFT^(Q)1x3 =
$$\left(\frac{107+W^{4X}}{12}\right)\left(\frac{107+W^{2X}}{12}\right)\left(\frac{107+W^{2}}{12}\right)\left(\frac{107+W^{2}}{12}\right)$$

FIRST FACTOR
NOTE WEST AN 8TH ROOT OF UNITY = COEFFICIENT
OF 117 IN TERM $\left(\frac{107+W^{4X}}{12}\right)$ CAN BE
DERIVED BY
W⁴ = W⁴(4xz+7x.+x)
W⁴ = W⁴(4xz+7x.+x)
W⁴ = W⁴(4xz+7x.+x)
U⁴ = U⁴(W⁴)^x = (-1)^x
(107+W⁴)^x = (-1)^x
To TE
FOR X0=0 $\left(\frac{107+U^{2}}{\sqrt{2}}\right)$
FUR XU=1 $\left(\frac{107-U^{7}}{\sqrt{2}}\right)$
THIS IS A KADAMARD TRAMSFORMATION

Middle (2nd) Factor for the 3 Qubit Example Construction

$$\frac{SECOND}{(MIDDLC)} \frac{FACTOR}{2(4x_{1}+2x_{1}+x_{n})} = W^{4x_{1}} W^{4x_{1$$

3rd (Least Significant) Factor for the 3 Qubit Example

$$\frac{THIRD}{FAKTOR} (LEAST SIGNIFICANT)$$

$$\frac{RICHTMOST}{RICHTMOST} = CATPUT FACTOR HAS AN W WITH NO
CERPONENT IN THE NUMERATOR
$$\frac{(4x_{k} + 2x_{1} + x_{n})}{(4x_{k} + 2x_{1} + x_{n})} = W^{4}x_{k} w^{2}x_{1} w^{2}$$

$$= (-1)^{x_{k}} (i)^{x_{1}} (w)^{x_{n}}$$

$$= (-1)^{x_{k}} (i)^{x_{1}} (w)^{x_{n}} (17)$$

$$\int \overline{\Sigma} = \int \frac{107 + (-1)^{x_{k}} (i)^{x_{1}} (17)}{\sqrt{\Sigma}}$$

$$\int \overline{\Sigma} = \frac{107 + (-1)^{x_{k}} (i)^{x_{1}} (17)}{\sqrt{\Sigma}}$$

$$\int \overline{\Sigma} = \frac{107 + (-1)^{x_{k}} (i)^{x_{1}} (w)^{x_{n}} (17)}{\sqrt{\Sigma}}$$$$

3rd (Least Significant) Factor for the 3 Qubit Example (cont'd)

$$\frac{T_{HIRO} \ F_{A \leq T \leq PR} \left(\circ B S G R V A TIONS\right)}{REDUKES TS}$$

$$when x_{n=0} \qquad THE \ (EAST \ SIGH IFICANT \ FACTOR \ REDUKES TS$$

$$x_{n=0} \qquad (\Delta + (-1)^{x_{2}} (c)^{x_{1}} | ID)$$

$$f_{2} \qquad f_{2} \qquad$$

3rd (Least Significant) Factor for the 3 Qubit Example (cont'd)



Building a Quantum Computing Circuit from the Mathematics of the Quantum Fourier Transform

QFT Quantum Gate Construction: 1st & 2nd Factor

BASED ON THE CALCULATION FROM FIRST FACTOR FIRST FACTOR (1x2) = H (x0) SECOND (MIDDLE) FACTOR 2 TERMS 1×27 = H 1×07 17:7 = { H 1xi7 No=0 R, H 1xi7 No=1 APPLY H T. THE TWO LEAST SIGNFFICANT KERS [XO) AND
 APPLY H T. THE TWO LEAST SIGNFFICANT KERS [XO) AND
 IX? UNCONDITIONALLY . H WILL ALWAYS BE USED
 IX? UNCONDITIONALLY . H WILL ALWAYS BE USED B CONDITIONALLY APPLY R. TO RESULT DF HIXIJIFXO=1 · APPLY TO LEAST SIGNIFICANT INPUT KETS 1X.7 1X.7 GET MOST STOPIFICANT PORTLON OF THE OUTPUTIS FASTARIZATION (SWAPPING NGEPS TO BE INCLUDED) RI 1207

QFT Quantum Gate Construction: 2nd Factor (cont'd)



QFT Quantum Gate Construction: 3rd Factor



QFT Quantum Gate Construction: 3rd Factor



Full Nth Order Qubit Quantum Fourier Transform Circuit



Last Slide